How small and how fast?

Shannon's Information Theoretic Limits for Data Compression and Communication

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Who is going to win the race?



- Who is going to win the race?
- "If you don't know where the stop line is, you cannot run the race effectively nor you can figure out how good you are."

Limits of communication and storage

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- Limits of communication and storage
- Compression : "What is the minimum length of a code which represents a given source losslessly?"

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... and more.



Digital Communication

Source Coding

Channel Coding



Source material



 "A Mathematical Theory of Communication" - Claude Shannon, Bell System Technical Journey, 1948

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Outline

Digital Communication

Source Coding

Channel Coding

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Communication System



- Clear what the receiver and the (analog) source is.
- What is the channel?

The Channel

- Given to us by nature (can optimise, but fundamental nature cannot be changed).
- Modelling noisy nature using probability (Models are not exact).
- Making appropriate assumptions are very important.
- AWGN : Typical Model for point to point (Noise signals are from a Gaussian Random Process)

The Digital Communication Model



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The Digital Communication Model



- Shannon did not explicitly deal with this model.
- Instead he merged together some of these blocks which helped him handle the problem probabilistically.

The Discrete Source



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The Discrete Channel



Shannon's Discrete (Digital) Communication Model



Source coding → Remove inherent redundancy in the source, represent in minimal fashion.

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Shannon's Discrete (Digital) Communication Model



Source coding → Remove inherent redundancy in the source, represent in minimal fashion.

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► Channel Coding → Add redundancy systematically to the source symbols to combat channel noise

Discrete Sources and Channels - Examples

Binary Source

Sampler + Quantize + Convert each quantized sample to bits.

Binary Symmetric Channel (BSC)

- Analog Channel is AWGN
- Assume BPSK Modulation and Threshold Detector at receiver.



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Source Coding

- \mathcal{X} Source Alphabet .
- ► X Source Random Variable.
- ► We assume that there is a probability distribution p(X) on the source.
- Want to compress this source store it in the least space without loss of information.

Binary Source

- $\blacktriangleright X \in \mathcal{X} = \{0,1\}$
- Source generates one binary symbol in each time unit.

▶
$$p(X = 0) = p$$
, $p(X = 1) = (1 - p)$. $(X \sim Ber(p))$

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- ▶ Need 1 bit to represent this source. Maximum length!

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- ▶ Need 1 bit to represent this source. Maximum length!
- Minimum length of representation per source symbol = Uncertainty/Randomness in the source
- ▶ What is the most random *M*-ary source? How many bits is required?
- What about arbitrary p for binary source?

Binary Entropy



To get the intermediate points, Shannon's idea was to 'let the source run' for some time.

Coin Toss - Bernouli RV

- Imagine a coin toss experiment with p(heads) = p, p(Tails) = 1 − p.
- Suppose we toss the coin N times with large N, how many heads and tails do we expect?

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- How many such N-length vectors are there?

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$$\blacktriangleright \left(\begin{array}{c} N\\ Np \end{array}\right) \text{ vectors.}$$

Coin Toss - Bernouli RV - Long vectors

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- ► $H(X) \triangleq -plog(p) (1 p)log(1 p)$ (Binary Entropy H(p))
- What is the probability of each such vector? **Ans**: $\approx 2^{-NH(X)}$.

• Holds with equality as $N \to \infty$.
Compressing a binary RV

Note that the source distribution begins to look like a uniform distribution for large N, with 2^{NH(X)} possible vectors, and Prob(any vector)≈ 2^{-NH(X)}.

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- ▶ No. of bits required to represent one source symbol $= \frac{NH(X)}{N} = H(X).$
- Using less number of bits than this will cause loss of information about source!

Shannon's source coding theorem

Shannon's Source Coding Theorem

The minimum number of bits required to represent a source random variable X taking values from \mathcal{X} with distribution p(X) is

$$H(X) = \sum_{x \in \mathcal{X}} -p(x) \log(p(x)).$$

An explicit scheme exists which can achieve compression arbitrarily close to H(X).

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- ► Channel is described according to conditional distribution $p(y|x), y \in \mathcal{Y}, x \in \mathcal{X}$. (Assume $p(y|x) = \prod_{i=1}^{N} p(y_i|x_i)$).



- Let Code alphabet be X with distribution p(X) (Codeword vector is x ∈ X^N).
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- Note that $p(y) = \sum_{x \in \mathcal{X}} p(y|x)p(x)$.
- Rate of the code $= \frac{K}{N}$ (no. of msg symbols per codeword symbol).
- Question: What is the maximum rate possible to achieve for (almost) zero probability of error?

Binary Source + Binary Symmetric Channel Setup Binary Symmetric Channel:



- Assume source symbols are binary and channel is BSC.
- Encoding: K-length binary strings to N-length binary strings.

• Number of codewords is 2^{K} .

Binary Source + Binary Symmetric Channel Setup Binary Symmetric Channel:



- Assume source symbols are binary and channel is BSC.
- Encoding: K-length binary strings to N-length binary strings.
- Number of codewords is 2^{K} .
- If X is channel input RV, we can also write channel output Y as

$$Y=X+Z,$$

where Z has a Ber(p) distribution and is independent of X.

Question: What is the best $\frac{K}{N} = \frac{\log_2(\text{No. of codewords})}{N}$ with (almost) zero probability of error?

▶ Repetition code: K = 1, mapped to N-length string with same bits as the input.

Probability of error goes to zero, what about rate?

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- Probability of error goes to zero, what about rate?
- Are there codes which have good rate, and low probability of error?
- Shannon proved such codes exist ! (Trade off with complexity of encoding/decoding)

- ▶ Let **x** be some *N*-length codeword input to the channel.
- Assuming large N, how does the received vector y look like?

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- ▶ Let *x* be some *N*-length codeword input to the channel.
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- ► Ans: We will have y = x + z, for some z containing roughly Np ones.

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- ▶ Let *x* be some *N*-length codeword input to the channel.
- Assuming large N, how does the received vector y look like?
- ► Ans: We will have y = x + z, for some z containing roughly Np ones.
- ▶ Number of such possible error vectors is $\begin{pmatrix} N \\ Np \end{pmatrix} \approx 2^{NH(Z)}$, and the probability of each is $\approx 2^{-NH(Z)}$.

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Non-intersecting Hamming spheres

- Sphere of 'radius' *Np* around *x*.
- > The received vector could be anywhere in this sphere.

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• Number of vectors in this sphere $\approx 2^{NH(Z)}$.

Non-intersecting Hamming spheres

- Sphere of 'radius' Np around x.
- ► The received vector could be anywhere in this sphere.
- Number of vectors in this sphere $\approx 2^{NH(Z)}$.
- Suppose there was another codeword x' such that the spheres around x and x' intersect.

Then there is decoding error

Any two codewords should be such that their spheres (of radius Np) don't intersect.

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- Pick the maximum possible number of such codewords following above rule. (Sphere packing)

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- Total possible received vectors $\approx 2^{NH(Y)}$.
- Number of vectors in each sphere $\approx 2^{NH(Z)}$.

- Any two codewords should be such that their spheres (of radius Np) don't intersect.
- Pick the maximum possible number of such codewords following above rule. (Sphere packing)
- Total possible received vectors $\approx 2^{NH(Y)}$.
- Number of vectors in each sphere $\approx 2^{NH(Z)}$.
- Hence Maximum number of codewords

$$\frac{2^{NH(Y)}}{2^{NH(Z)}} = 2^{N(H(Y) - H(Z))}.$$

• Maximum rate of code (with output distribution p(y)) = H(Y) - H(Z).

Note that H(Y) is a function of p(y) which is a function of p(x).

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- Note that H(Y) is a function of p(y) which is a function of p(x).
- The maximum rate of transmission through the BSC is thus

$$max_{p(x)}(H(Y) - H(Z))$$

= $max_{p(x)}H(Y) - H(p)$
= $1 - H(p)$

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This is called the Capacity of the BSC.



Channel Capacity Theorem

For any discrete-memoryless channel with given p(y|x), the rate of transmission R is always $\leq C$, where C is the channel capacity given as

$$C \triangleq max_{p(x)}H(Y) - H(Y|X).$$

Also, there exists some encoding scheme by which any rate arbitrarily close to capacity is achievable.

Capacity of AWGN

- For AWGN with bandwidth W:: Capacity = $\frac{1}{2}log(1 + SNR) = \frac{1}{2}log(1 + \frac{P}{N_0W}).$
- Only Existence of Good Codes is shown by Shannon.

 Construction of 'good' codes has happened (for AWGN channels) over the last several decades since Shannon.

Capacity curve for AWGN



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Turbo Codes and LDPC Codes along Shannon Capacity Curve

Power Efficiency of Standard Binary Channel Codes



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- Who is going to win the race?
- "If you don't know where the stop line is, you cannot run the race effectively nor you can figure out how good you are."

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- Who is going to win the race?
- "If you don't know where the stop line is, you cannot run the race effectively nor you can figure out how good you are."
- Shannon tells us where the stop line is!

Thank You!